A Psychoacoustic "NofM"-Type Speech Coding Strategy for Cochlear Implants

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We describe a new signal processing technique for cochlear implants using a psychoacoustic-masking model. The technique is based on the principle of a so-called "NofM" strategy. These strategies stimulate fewer channels (N) per cycle than active electrodes (NofM; N < M). In "NofM" strategies such as ACE or SPEAK, only the N channels with higher amplitudes are stimulated. The new strategy is based on the ACE strategy but uses a psychoacoustic-masking model in order to determine the essential components of any given audio signal. This new strategy was tested on device users in an acute study, with either 4 or 8 channels stimulated per cycle. For the first condition (4 channels), the mean improvement over the ACE strategy was 17%. For the second condition (8 channels), no significant difference was found between the two strategies.

Keywords and phrases: cochlear implant, NofM, ACE, speech coding, psychoacoustic model, masking.

1. INTRODUCTION

Cochlear implants are widely accepted as the most effective means of improving the auditory receptive abilities of people with profound hearing loss. Generally, these devices consist of a microphone, a speech processor, a transmitter, a receiver, and an electrode array which is positioned inside the cochlea. The speech processor is responsible for decomposing the input audio signal into different frequency bands or channels and delivering the most appropriate stimulation pattern to the electrodes. When signal processing strategies like continuous interleaved sampling (CIS) [1] or advanced combinational encoder (ACE) [2, 3, 4] are used, electrodes near the base of the cochlea represent high-frequency information, whereas those near to the apex transmit low-frequency information. A more detailed description of the process by which the audio signal is converted into electrical stimuli is given in [5].

Speech coding strategies play an extremely important role in maximizing the user's overall communicative potential, and different speech processing strategies have been developed over the past two decades to mimic firing patterns inside the cochlea as naturally as possible [5]. "NofM" strategies such as ACE or spectral peak (SPEAK) [4] were developed in the 1990s. These strategies separate speech signals into M subbands and derive envelope information from each band signal. N bands with the largest amplitude are then selected for stimulation (N out of M). The basic aim here is to increase the temporal resolution by neglecting the less significant spectral components and to concentrate on the more important features. These strategies have demonstrated either a significant improvement or at least

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FIGURE 1: Block diagram illustrating ACE.

user preference over conventional CIS-like strategies [6, 7, 8]. However, speech recognition for cochlear implant recipients in noisy conditions—and, for some individuals, even in quiet—remains a challenge [9, 10]. To further improve speech perception in cochlear implant users, the authors decided to modify the channel selection algorithm of the ACE speech coding strategy.

This work therefore describes a new method for selecting the N bands used in "NofM" strategies. As outlined above, conventional "NofM" strategies select the N bands with the largest amplitudes from the M filter outputs of the filter bank. In the new scheme the N bands are chosen using a psychoacoustic-masking model. The basic structure of this strategy is based on the ACE strategy but incorporating the above-mentioned psychoacoustic model. This new strategy has been named the psychoacoustic advanced combination encoder (PACE). Psychoacoustic-masking models are derived from psychoacoustic measurements conducted on normal-hearing persons [11, 12, 13] and can be used to extract the most meaningful components of any given audio signal [14, 15]. Those techniques are widely used in common hi-fi data reduction algorithms, where data streams have to be reduced owing to bandwidth or capacity limitations. Wellknown examples of these techniques are the adaptive transform acoustic coding (ATRAC) [16] coding system for minidisc recorders and the MP3 [17, 18] compression algorithm for transferring music via the Internet. These algorithms are able to reduce the data to one-tenth of its original volume with no noticeable loss of sound quality.

"NofM" speech coding strategies have some similarities to the above-mentioned hi-fi data reduction or compression algorithms in that these strategies also compress the audio signals by selecting only a subset of the frequency bands. The aim in introducing a psychoacoustic model for channel selection was to achieve more natural sound reproduction in cochlear implant users.

Standardized speech intelligibility tests were conducted using both the ACE and the new PACE strategy, and the scores compared in order to test whether the use of a psychoacoustic model in the field of cochlear implant speech coding can indeed yield improved speech understanding in the users of these devices. The paper is organized as follows. In Section 2, a review of the ACE strategy is presented. Furthermore, the psychoacoustic model and how it has been incorporated into an "NofM" strategy is described. Section 3 gives the results of the speech understanding tests with cochlear implant users and finally, in Sections 4 and 5, a discussion and the conclusions are presented respectively.

2. METHODS

2.1. Review of the ACE strategy

Several speech processing strategies have been developed over the years. These strategies can be classified into two groups: those based on feature extraction of the speech signals and those based on waveform representation. The advanced combinational encoder (ACE) [2, 3] strategy used with the Nucleus implant is an "NofM"-type strategy belonging to the second group. The spectral peak (SPEAK) [4] strategy is identical in many aspects to the ACE strategy, but different in rate. Figure 1 shows the basic block diagram illustrating the ACE strategy.

The signal from the microphone is first pre-emphasized by a filter that amplifies the high-frequency components in particular. Adaptive-gain control (AGC) is then used to limit distortion of loud sounds by reducing the amplification at the right time.

Afterwards, the signal is digitized and sent through a filter bank. ACE does not explicitly define a certain filter bank approach. The frequency bounds of the filter bank are linearly spaced below 1000 Hz, and logarithmically spaced above 1000 Hz.

An estimation of the envelope is calculated for each spectral band of the audio signal. The envelopes are obtained by computing the magnitude of the complex output. Each band pass filter is allocated to one electrode and represents one channel. For each frame of the audio signal, *N* electrodes are stimulated sequentially and one cycle of stimulation is completed. The number of cycles/second thus determines the rate of stimulation on a single channel, also known as channel stimulation rate.



FIGURE 2: Block diagram illustrating research ACE.

TABLE 1: Number of FFT bins, center frequencies, and gains per filter band for M = 22.

| Band number <i>z</i> | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 |
|----------------------|------|------|------|------|------|------|------|------|------|------|------|
| Number of bins | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 2 | 2 |
| Center freqs. (Hz) | 250 | 375 | 500 | 625 | 750 | 875 | 1000 | 1125 | 1250 | 1437 | 1687 |
| Gains g_z | 0.98 | 0.98 | 0.98 | 0.98 | 0.98 | 0.98 | 0.98 | 0.98 | 0.98 | 0.68 | 0.68 |
| | • | | | | | | | | | | |
| Band number <i>z</i> | 12 | 13 | 14 | 15 | 16 | 17 | 18 | 19 | 20 | 21 | 22 |
| Number of bins | 2 | 2 | 3 | 3 | 4 | 4 | 5 | 5 | 6 | 7 | 8 |
| Center freqs. (Hz) | 1937 | 2187 | 2500 | 2875 | 3312 | 3812 | 4375 | 5000 | 5687 | 6500 | 7437 |
| Gains g_z | 0.68 | 0.68 | 0.65 | 0.65 | 0.65 | 0.65 | 0.65 | 0.65 | 0.65 | 0.65 | 0.65 |

The bandwidth of a cochlear implant is limited by the number of channels (electrodes) and the overall stimulation rate. The channel stimulation rate represents the temporal resolution of the implant, while the total number of electrodes M represents the frequency resolution. However, only N out of M electrodes (N < M) are stimulated in each cycle, therefore a subset of filter bank output samples with the largest amplitude is selected. If N is decreased, the spectral representation of the audio signal becomes poorer, but the channel stimulation rate can be increased, giving a better temporal representation of the audio signal. Conversely, if the channel stimulation rate is decreased, N can be increased, giving a better spectral representation of the audio signal.

Finally, the last stage of the process maps the amplitudes to the corresponding electrodes, compressing the acoustic amplitudes into the subject's dynamic range between measured threshold and maximum comfortable loudness level for electrical stimulation.

2.2. Research ACE strategy used

A research ACE strategy [3] was made available by Cochlear Corporation for the purpose of deriving new speech coding strategies. However, the research ACE strategy is designed to process signals that are already digitized. For this reason, the pre-emphasis filter and adaptive-gain controls (AGC) incorporated at the analogue stage are not included in this set-up. Figure 2 shows a basic block diagram illustrating the strategy.

A digital signal sampled at 16 kHz is sent through a filter bank without either pre-amplification or adaptive-gain control. The filter bank is implemented with an FFT (fast Fourier transform). The block update rate of the FFT is adapted to the rate of stimulation on a channel (i.e., the total implant rate divided by the number of bands selected N). The FFT is performed on input blocks of 128 samples (L = 128) of the previously windowed audio signal. The window used is a 128-point Hann window [19]

$$w(j) = 0.5 \left(1.0 - \cos\left(\frac{2\pi j}{L}\right) \right), \quad j = 0, \dots, L - 1.$$
 (1)

The linearly-spaced FFT bins are then combined by summing the powers to provide the required number of frequency bands M, thus obtaining the envelope in each spectral band a(z) (z = 1, ..., M). The real part of the *j*th FFT bin is denoted with x(j), and the imaginary part y(j). The power of the bin is

$$r^{2}(j) = x^{2}(j) + y^{2}(j), \quad j = 0, \dots, L - 1.$$
 (2)

The power of the envelope of a filter band z is calculated as a weighted sum of the FFT bin powers

$$a^{2}(z) = \sum_{j=0}^{L/2} g_{z}(j)r^{2}(j), \quad z = 1, \dots, M,$$
(3)

where $g_z(j)$ are set to the gains g_z for a specific number of bins and otherwise zero. This mapping is specified by the number of bins, selected in ascending order starting at bin 2, and by the gains g_z as presented in Table 1 [3, 20].

The envelope of the filter band *z* is

$$a(z) = \sqrt{\sum_{j=0}^{L/2} g_z(j) r^2(j)}, \quad z = 1, \dots, M.$$
(4)

In the "sampling and selection" block, a subset of N (N < M) filter bank envelopes $a(z_i)$ with the largest amplitude are selected for stimulation.



FIGURE 3: Block diagram illustrating an "NofM" strategy incorporating a psychoacoustic model for selecting the *N* bands. The strategy may be termed the psychoacoustic ACE strategy.

The "mapping" block, determines the current level from the envelope magnitude and the channel characteristics. This is done by using the loudness growth function (LGF) which is a logarithmically-shaped function that maps the acoustic envelope amplitude $a(z_i)$ to an electrical magnitude

$$p(z_i) = \begin{cases} \frac{\log(1 + \rho((a(z_i) - s)/(m - s)))}{\log(1 + \rho)}, & s \le a(z_i) \le m, \\ 0, & a(z_i) < s, \\ 1, & a(z_i) \ge m. \end{cases}$$
(5)

The magnitude $p(z_i)$ is a fraction in the range 0 to 1 that represents the proportion of the output range (from the threshold *T* to the comfort level *C*). A description of the process by which the audio signal is converted into electrical stimuli is given in [21]. An input at the base-level *s* is mapped to an output at threshold level, and no output is produced for an input of lower amplitude. The parameter *m* is the input level at which the output saturates; inputs at this level or above result in stimuli at comfort level. If there are less than *N* envelopes above base level, they are mapped to the threshold level. The parameter ρ controls the steepness of the LGF, the selection of a suitable value for ρ is described in [20].

Finally, the channels z_i , are stimulated sequentially with a stimulation order from high-to-low frequencies (base-to-apex) with levels:

$$l_i = T + (C - T)p_i.$$
 (6)

2.3. "NofM" strategy using a psychoacoustic model: the psychoacoustic ACE (PACE) strategy

Based on the general structure of the research ACE strategy (Figure 2) but incorporating a psychoacoustic model, a new approach was designed in order to select the N (N < M) bands in "NofM" strategies. A basic block diagram illustrating the proposed PACE strategy is presented in Figure 3.

Both the filter bank and the envelope detection process are identical to those in the research ACE strategy. A psychoacoustic-masking model—as opposed to a peakpicking algorithm—is then used to select the N bands. Consequently, the bands selected by this new approach are not necessarily those with the largest amplitudes (as is the case in the ACE strategy) but the ones that are, in terms of hearing perception, most important to normal-hearing people. Afterwards, the bands selected are mapped to electrical impulses and sent to the electrode array following exactly the same process as in the research ACE strategy.

In the following paragraphs the psychoacoustic model and the selection algorithm will be explained.

2.3.1. Psychoacoustic model

There are different classes of psychoacoustic models, the one referred to in this manuscript being a psychoacousticmasking model. Such models describe masking effects that take place in a healthy auditory system. Psychoacoustic models have been successfully used within the field of audio coding in order to reduce bandwidth requirements by removing the less perceptually important components of audio signals. Because "NofM" speech coding strategies only select certain spectral elements of the audio signals, it can be speculated that a psychoacoustic model may ensure more effective selection of the most relevant bands than is achieved by merely selecting the spectral maxima, as with the ACE strategy.

Psychoacoustic-masking models are based on numerous studies of human perception, including investigations on the absolute threshold of hearing and simultaneous masking. These effects have been studied by various authors [11, 12, 13, 22].

The absolute threshold of hearing is a function that gives the required sound pressure level (SPL) needed in order that a pure tone is audible in a noiseless environment. The effect of simultaneous masking occurs when one sound makes it difficult or impossible to perceive another sound of similar frequency.

A psychoacoustic model as described by Baumgarte in 1995 [15] was adapted to the features of the ACE strategy. The psychoacoustic model employed here is used to select the N most significant bands in each stimulation cycle. In the following sections we describe the steps (shown in Figure 4) that constitute the masking model. The masked threshold is calculated individually for each band selected. The overall masked threshold created by the different bands can then be approximated by nonlinear superposition of the particular masked thresholds. Figure 4 shows an example of the psychoacoustic model implemented operating on two selected bands.



FIGURE 4: (a) Block diagram. The input comprises the envelope values of the bands chosen by the selection algorithm. The output is the overall masked threshold. (b) Associated levels over the frequency band number z.



FIGURE 5: (a) Threshold in quiet over the frequency in Hz. (b) Threshold in quiet approximation over the band number *z* and spectral level when the vowel "A" is uttered.

2.3.1.1. Threshold in quiet

A typical absolute threshold expressed in terms of dB SPL is presented in Figure 5a [23].

The function $L_{abs(z)}$ representing the threshold in quiet in each frequency band z is obtained by choosing one representative value of the function presented in Figure 5a at the centre frequency of each frequency band (Table 1). However, as the authors have no a priori knowledge regarding playback levels (SPL) of the original audio signals, a reference had to be chosen for setting the level of the threshold in quiet. It is known that the threshold in quiet lies at around 50 dB below "normal speech level" (i.e., between 200 Hz and 6 kHz [11]). The level of the function $L_{abs}(z)$ was therefore set at 50 dB below the level of the voiced parts from certain audio samples used as test material. Figure 5b presents the resulting $L_{abs}(z)$ and the spectral level obtained when a generic vowel "a" in the test material is uttered. The vowel "a" was stored in a "wav" file format coded with 16 bits per sample, and the standard deviation for the whole vowel was about 12 dB below the maximum possible output level. It is important to note that $T_{abs}(f)$ is expressed in terms of dB SPL and $L_{abs}(z)$ in dB (0 dB corresponds to the minimum value of the threshold in quiet mentioned before).

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FIGURE 6: Spreading function $L_i(z)$ of one masker component $A(z_i)$ at the band z_i . The left and right slopes of the spreading function are indicated as s_i and s_r . The attenuation of the maximum relative to the masker level is denoted by a_v .

2.3.1.2. Masking pattern of single stimulating component

For each selected band, a function is calculated that models the masking effect of this band upon the others. This function familiar in the field of psychoacoustics as the so-called spreading function, expressed with the same dB units as in Figure 5b, is presented in Figure 6.

The spreading function is described by three parameters: attenuation, left slope, and right slope. The amplitude of the spreading function is defined using the attenuation parameter a_v . This parameter is defined as the difference between the amplitude of the selected band $A(z_i)$ and the maximum of the spreading function in dB units. The slopes s_l and s_r correspond to the left and right slopes, respectively, in the unit "dB/band." As presented in [15], the spreading function belonging to a band z_i with amplitude $A(z_i)$ in decibels is mathematically represented by $L_i(z)$:

$$L_{i}(z) = \begin{cases} A(z_{i}) - a_{v} - s_{l} \cdot (z_{i} - z), & z < z_{i}, \\ A(z_{i}) - a_{v} - s_{r} \cdot (z - z_{i}), & z \ge z_{i}, \end{cases}$$
(7)

where

- (i) *z* denotes the frequency band number at the output of the filter bank, 1 ≤ *z* ≤ *M*,
- (ii) *i* denotes that the band selected is z_i (i.e., masker band).

In the model description of [15], z denoted the critical band rate [11, 24] or equivalently critical band number [12, 13]. Because the bandwidths of the frequency bands used in the filter bank in the ACE and PACE schemes are approximately equal to the critical bands, the frequency band number corresponds approximately to the critical band rate. Therefore, in the implementation of the masking model in the present study, it was opted to define the masking patterns as a function of the frequency band number instead of the critical band rate.

2.3.1.3. Nonlinear superposition

The sound intensities $I_{abs}(z)$ and $I_i(z)$ are calculated from the decibel levels by

$$I_{abs}(z) = 10^{L_{abs}(z)/10},$$

$$I_i(z) = 10^{L_i(z)/10}.$$
(8)

Threshold components should be combined in a way that reflects the characteristics of human auditory perception. Certain approaches have been based on linear addition of the threshold components [25]. However, further results proved that linear models fail in most cases where threshold components exhibit spectral overlapping [25, 26]. A nonlinear model was thus proposed to reproduce the significantly higher masking effects obtained in the overlapping threshold components by linear models [27]. Differences of the masked thresholds resulting from a linear and nonlinear superposition are discussed in [15]. Results indicate that significant improvements are possible using a nonlinear model.

A "power-law model," as described in 1995 by Baumgarte [15], was therefore used for the superposition of different masked thresholds in order to represent the nonlinear superposition. The "power-law model" is defined by the parameter α where $0 < \alpha \le 1$. If α is 1, the superposition of thresholds is linear; if α is lower than 1, the superposition is carried out in a nonlinear mode. A description of different values of α can be also obtained from [15]. The nonlinear superposition of masking thresholds defined by $I_T(z)$ is

$$I_T(z) = \left[\left[I_{abs}(z) \right]^{\alpha} + \sum_i \left[I_i(z) \right]^{\alpha} \right]^{1/\alpha}.$$
 (9)

The level in decibels of the superposition of the individual masking thresholds denoted by $L_T(z)$ is

$$L_T(z) = 10 \log_{10} (I_T(z)).$$
(10)



FIGURE 7: Selection algorithm: the audio samples are the input and the *N* bands selected are the output. A psychoacoustic model is used to select the bands in each iteration.

2.3.2. Selection algorithm

This algorithm is inspired by the analysis/synthesis loop [14] used in the MPEG-4 parametric audio coding tools "harmonic and individual lines plus noise" (HILN) [28]. The selection algorithm loop chooses the *N* bands iteratively in order of their "significance" (Figure 7).

The amplitude envelopes of the *M* bands A(z) (z = 1,...,M) are obtained from the filter bank. For the first iteration of the algorithm there is no masking threshold and the threshold in quiet is not considered; the first band selected is therefore the one with the largest amplitude. For this band, the psychoacoustic model calculates its associated masking threshold $L_T(z)$ (z = 1,...,M).

In the next iteration the band z_i is selected out of the remaining M - 1 bands for which the following difference is largest:

$$z_i = \operatorname{argmax} (A(z) - L_T(z)), \quad z = 1, \dots, M.$$
 (11)

The individual masking threshold of this band $L_i(z)$ is calculated and added to the one previously determined. The masking threshold $L_T(z)$ for the actual iteration is then obtained and used to select the following band. The loop (Figure 7) is repeated until the *N* bands are selected. Therefore, at each step of the loop, the psychoacoustic model selects the band that is considered as most significant in terms of perception.

2.3.3. Application to the ACE strategy

The psychoacoustic model has been incorporated into a research ACE strategy made available by Cochlear Corporation as a Matlab "toolbox," designated the nucleus implant communicator (NIC). However, this ACE strategy does not incorporate the pre-emphasis and adaptive-gain control filters described in Section 2.1. The new strategy based on psychoacoustic masking has been termed the psychoacoustic ACE (PACE) strategy as explained in Section 2.3. The NIC allows the ACE and the PACE to be configured using different parameters: the rate of stimulation on a channel (channel stimulation rate), the number of electrodes or channels into which the audio signal is decomposed (M), and the number of bands selected per cycle (N). At the same time, the psychoacoustic model can be modified according to the parameters that define the spreading function (Figure 6). In the following paragraphs we will describe the rationale for setting the parameter values that are used in the experiments.

2.3.3.1. Parameter setting for the PACE strategy

The parameter set that defines the spreading function should describe the spectral masking effects that take place in a healthy auditory system. Such effects depend strongly on the type of components that are masking and being masked [11]. However, they can be reduced to two general situations: masking of pure tones by noise and masking of pure tones by tones [11]. Furthermore, the first scenario should identify the type of masking noise, that is, whether it is broadband, narrowband, lowpass or highpass noise. For the second scenario, it should also be specified which kind of tone is having a masking effect, that is, whether it is pure tone or a set of complex tones. For each of these situations a different parameter set for the spreading function should be defined, depending on the frequencies and amplitudes of the masker and masked components. For example, in audio compression algorithms such as the MPEG1 layer 3 (MP3) [17] usually only two situations are considered [23]: noisemasking tone (NMT) and tone-masking noise (TMN). For each scenario, a different shape for the spreading function based on empirical results is defined.

The psychoacoustic model applied in this pilot study does not discriminate between tonal and noise components. Furthermore, it is difficult to specify a set of parameters for the spreading function based on empirical results as with the MP3. The parameters of the spreading function in the MP3 can be set through empirical results with normal hearing people. There are a lot of studies in this field which can be used to set the parameters of the spreading function in all the situations mentioned before. However, with cochlear implant users there is relatively little data in this field. For this reason, the results of previous studies by different authors with normal hearing people [11, 12, 13] were incorporated into a unique spreading function approximating all the masking situations discussed above. In these studies the necessity became apparent for the right slope of the spreading function to be less steep than the left slope. In consequence, the left slope of the PACE psychoacoustic model was always set to higher dB/band values than the right slope. Two configurations for the left and right slopes were chosen in order to



FIGURE 8: (a) Frequency band decomposition of one frame coming from a token of the vowel "a." (b) Selected bands using the ACE strategy for one frame coming from a token of the vowel "a."

test different masking effects: (left slope = 12 dB/band, right slope = 7 dB/band) and (left slope = 40 dB/band, right slope = 30 dB/band). Furthermore, outcomes from previous studies demonstrated that the value of a_v defining the attenuation of the spreading function with regard to the masker level is highly variable, ranging between 4 dB and 24 dB depending on the type of masker component [23]. For this reason, the value of a_v was set to 10 dB, which lies between the values mentioned above. The parameter α which controls the non-linear superposition of individual masking thresholds was set to 0.25, which is in the range of values proposed in [15, 27]. Finally, the threshold in quiet was set to an appropriate level as presented in Section 2.3.1.1.

2.3.3.2. Objective analysis

The NIC software described permits a comparison between the ACE strategy and the psychoacoustic ACE strategy. Figure 8a shows the frequency decomposition of a speech token processed with both strategies. The token is the vowel introduced in Section 2.3.1.1. The filter bank used for both strategies decomposes the audio signal into 22 bands (M =22). Eight of the separated-out bands are selected (N = 8). The bands selected differ between the two strategies, as different methods of selecting the amplitudes were used. Figure 8b gives the bands selected by the ACE strategy. Figures 9a, 9b, 10a, and 10b, respectively, illustrate the bands selected by the PACE strategy and the spreading functions used in the psychoacoustic model.

The spreading function presented in Figure 10b is steeper than that demonstrated in Figure 9b. Thus, using the psychoacoustic model based on the spreading function in Figure 9b, any frequency band will have a stronger masking effect over the adjacent frequency bands than with the psychoacoustic model based on the spreading function in Figure 10b. The psychoacoustic models based on the spreading function shown in Figures 9b and 10b are referred to in the following sections as psychoacoustic models 1 and 2, respectively.

Looking at Figures 8, 9, and 10 it can be observed that the bands selected using a psychoacoustic model are distributed broadly across the frequency range, in contrast to the stimulation pattern obtained with the simple peakpicking "NofM" approach used in the standard ACE strategy. The ACE strategy tends to select groups of consecutive frequency bands, increasing the likelihood of channel interaction between adjacent electrodes inside the cochlea. In the PACE strategy, however, the selection of clusters is avoided owing to the masking effect that is exploited in the psychoacoustic model. This feature can be confirmed by an experiment that involves counting the number of clusters of different lengths selected by the ACE and PACE strategies during the presentation of 50 sentences from a standardized sentence test [29]. For the PACE the test material was processed twice, the first time using psychoacoustic model 1 and then using psychoacoustic model 2. The 50 sentences were processed using a channel stimulation rate of 500 Hz and selecting 8 bands in each frame for both strategies. This means that the maximum possible cluster length is 8, when all selected bands are sequenced consecutively across the frequency range as demonstrated in Figure 8b. The minimum possible cluster length is 1, which occurs when all selected bands are separated from each other by at least one channel. Table 2 presents the number of clusters of different lengths (1–8) for the ACE, PACE 1 (using psychoacoustic model 1) and PACE 2 (using psychoacoustic model 2) strategies that occur during the 50 sample sentences.

The data clearly show that ACE tends on average to produce longer clusters than PACE 1 or PACE 2. At cluster length eight, for example, the ACE strategy selects 3607 clusters,



FIGURE 9: (a) Selected bands using the PACE strategy for one frame coming from a token of the vowel "a." (b) Spreading function used in the psychoacoustic model (left slope = 12 dB/band, right slope = 7 dB/band, $a_v = 10$ dB).



FIGURE 10: (a) Selected bands using the PACE strategy for one frame coming from a token of the vowel "a." (b) Spreading function used in the psychoacoustic model (left slope = 40 dB/band, right slope = 30 dB/band, $a_v = 10$ dB).

whereas the PACE strategy with the psychoacoustic model 1 selects only 33 and the PACE strategy with the psychoacoustic model 2 selects 405. The fact that the PACE 1 selects fewer clusters of 8 bands than the PACE 2 is attributable to the masking effect of the first psychoacoustic model being stronger than the second, as defined by the spreading functions of Figures 9b and 10b.

2.4. Speech intelligibility tests

2.4.1. Test environment

The strategies programmed within the NIC environment were tested with patients using a Nucleus 24 implant manufactured by Cochlear Corporation. The NIC software permits the researcher to communicate with the Nucleus implant and to send any stimulus pattern to any of the 22 electrodes. The NIC communicates with the implant via the standard hardware also used for fitting recipients in routine clinical practice. A specially initialized clinical speech processor serves as a transmitter for the instructions from the personal computer (PC) to the subject's implant (Figure 11), so that the clinical processor does not itself perform any speech coding computations. The NIC, in conjunction with Matlab, processes the audio signals on a PC. An interface then provides the necessary functionality for a user application that takes signals, processed using the Matlab toolbox, and transmits them to the cochlear implant via the above-mentioned speech processor.

TABLE 2: Number of times that consecutive frequency bands or clusters are selected for different group lengths for the ACE and PACE strategies (using psychoacoustic model 1) and PACE (using psychoacoustic model 2).

| Cluster | Number of | Number of | Number of |
|---------|--------------|-----------------|-----------------|
| length | ACE clusters | PACE 1 clusters | PACE 2 clusters |
| 1 | 60 564 | 370 161 | 186 338 |
| 2 | 34 248 | 107 057 | 114 201 |
| 3 | 20 557 | 21 449 | 46 124 |
| 4 | 15 382 | 3509 | 18314 |
| 5 | 12671 | 1424 | 8356 |
| 6 | 15 287 | 943 | 3129 |
| 7 | 17 153 | 566 | 1382 |
| 8 | 3607 | 33 | 405 |

The Nucleus 24 implant can use up to a maximum of 22 electrodes. However, only 20 electrodes were used by all of our test subjects as their speech processor in everyday use, the "ESPrit 3G," only supports 20 channels and the testees were accustomed to that configuration. For this reason, the two most basal channels were dropped from the original filter bank presented in Section 2.2 and thus could not be selected for stimulation.

2.4.2. Subjects

Eight adult users of the Nucleus 22 cochlear implant system participated in this study. The relevant details for all subjects are presented in Table 3. All test subjects used the ACE strategy in daily life and all were at least able to understand speech in quiet.

2.4.3. Study design

The test material used was the HSM (Hochmair, Schulz, Moser) sentence test [29]. Together with the Oldenburger sentence test [30], this German sentence test is well accepted among German CI centres as a measure of speech perception in cochlear implant subjects. It consists of 30 lists, each with a total of 106 words in 20 everyday sentences consisting of three to eight words. Scoring is based on "words correct." The test was created to minimize outcome variations between the lists. A study involving 16 normal-hearing subjects in noisy conditions (SNR = -10 dB) yielded 51.3% correctly repeated words from the lists, with a small range of only 49.8% to 52.6% [29]. The test can be administered in quiet and noise. The noise has a speech-shaped spectrum as standardized in CCITT Rec. 227 [31], and is added keeping fixed the overall output level of the test material.

In order to find suitable parameters of the spreading function in the PACE strategy, HSM test material was processed using two different parameter settings for the spreading function, as described in Section 2.3.3.1. Test signals were then delivered to the implants and the subjects reported which samples sounded clearer and more comfortable. The signals were presented in both quiet and noise. The channel stimulation rate was adapted to the needs of each user and



FIGURE 11: Research hardware made available by cochlear corporation.

both 4 and 8 maxima were tried. This procedure was carried out on 3 subjects over a period of several hours. All 3 subjects reported that the sound was best when using the spreading function shown in Figure 10b (psychoacoustic model 2). This particular spreading function was subsequently used for all 8 test subjects listed in Table 3.

All tests had to be conducted on an acute basis as the described research environment does not permit any chronic use, that is, take home experience. In generating the subject's program, the same psychophysical data measured in the R126 clinical fitting software were used in both the ACE and PACE programs. The parameters that define the loudness growth function (see Section 2.2): the base level of the loudness S, the saturation level M, and the steepness parameter ρ were set for all the patients to 33.86 dB, 65.35 dB, and 416.2063, respectively, which are the default parameters in the clinical fitting software [2, 20]. However, the S and M values were converted to the linear amplitudes s and m in order to be inserted in (5) according to the scaling described in Section 2.3.1. Using these values guaranteed that the level of the HSM sentence test was correctly mapped into the dynamic range defined by S and M. The threshold and maximum comfortable levels were adjusted to the needs of each patient. Before commencing actual testing, some sample sentences were processed using both the ACE and PACE strategies. The test subjects spent some minutes listening to the processed material, using both strategies, in order to become familiarized with them. At the same time, the volume was adjusted to suit the needs of the subjects by increasing or decreasing the value of the comfort and threshold levels.

For the actual testing, at least 2 lists of 20 sentences were presented in each condition, with the same number of lists used for both the ACE and PACE conditions. Sentences were presented either in quiet or in noise, depending on the subject's performance (Table 4). The lists of sentences were processed by the ACE and PACE strategies, with either 4 or 8 bands selected per frame. The order of the lists

| Patient id | Age | Duration of deafness (y) | Cause of deafness | Implant experience (Nucleus 24) (y) | Electrode type | Usual strategy |
|------------|------------|-----------------------------|-------------------|----------------------------------------|----------------|----------------|
| P1 | 65 | 0.75 | Unknown | 8 | Straight | ACE 8 of 20 |
| 11 | 05 | 0.75 | Clikilowii | 0 | oungit | 1080 pps |
| D2 | 42 | 0.08 | Temporal | 4 | Straight | ACE 8 of 20 |
| 12 | 42 | 0.00 | bone fracture | Ŧ | otraight | 1080 pps |
| D3 | 67 | 11 | Otosclarasis | 1 | Contour | ACE 8 of 20 |
| 15 | 07 | 11 | Otoscierosis | 1 | Contour | 900 pps |
| D4 | 30 | 1 21 | Unknown | 6 | Straight | ACE 8 of 20 |
| 14 | 57 | 1.21 | UIKIIOWII | 0 | otraight | 500 pps |
| D 5 | 19 | 216 | Sudden | 15 | Contour | ACE 8 of 20 |
| 15 | 49 2.10 | 2.10 | hearing loss | 1.5 | Contour | 900 pps |
| P6 | 64 | 0.66 | Infection | 0.6 | Contour | ACE 8 of 20 |
| 10 | 10 04 0.00 | 0.00 | meetion | 0.0 | Contour | 900 pps |
| P 7 | 68 | 13 | Unknown | 2 | Contour | ACE 8 of 20 |
| 17 | 00 | 1.5 | CHKHOWH | 2 | Contour | 1200 pps |
| P8 | 55 | 0.08 | Sudden | 0.08 | Contour | ACE 8 of 20 |
| 10 | Po 55 0.08 | | hearing loss | 0.00 | Contour | 900 pps |

TABLE 3: Subject demographics.

was randomized and the subjects had to repeat each sentence without knowing which strategy they were listening to (ACE or PACE).

As both strategies were tested on the same hardware and are based on the same psychophysical parameters, the tests permitted a fair comparison.

3. RESULTS

All subjects reported that the sound experienced using both strategies was understandable and not very different from what they were used to hearing through their everyday ACE strategy. Subjects 4 and 8 were only presented with sentences in quiet, as they were unable to understand speech in noise. Subject 1 reported that he could not perceive any difference between the two strategies. The other 7 subjects reported that the auditory sensations perceived using the new strategy were more melodious and clearer than those with the ACE, although the everyday speech-coding strategy used by their clinical speech processors was ACE. Subjects 6 and 8 had the impression that the person talking spoke more rapidly when using ACE—a common finding when cochlear implant users are having difficulties in understanding the test material.

Figures 12 and 13 present the averaged scores obtained by each subject for the different tests performed under two conditions, that is, stimulating either 4 or 8 of a total of 20 channels in each cycle. The tests were carried out in noise, with a signal-to-noise ratio of 15 dB (unless otherwise stated).

The results obtained show that all 8 subjects obtained better or equal scores using the PACE strategy when 4 electrodes were stimulated in each frame. When 8 electrodes were stimulated, only subject 7 obtained a better score using the ACE strategy than with the PACE strategy. Subject 2 achieved a better result using 4 electrodes with PACE than when using 8 electrodes with the ACE strategy. However, this may be due to a degree of variability within the test material or simply because of the subjects' diminished concentration at the end of the test session.

The scores show that the difference between the averaged groups becomes more marked when 4 electrodes are selected in each cycle instead of 8. In the former case, as fewer electrodes are stimulated, it becomes more important to select the most relevant amplitudes for each cycle. It was also observed that, when using PACE, performance using 4 electrodes matched that achieved with 8. That indicates that PACE may be able to generate the same scores as ACE while using only half as many electrodes. No significant difference could be found between the 8-channel ACE and 8-channel PACE condition. The above results are supported by the statistical analysis described below.

The program used for the analysis was SPSS V 12; the results were subjected to the Wilcoxon test [32]. Table 5 shows the outcome of the statistical analysis.

The statistical results show that the PACE strategy was found to yield a significant advantage only when 4 channels were selected for stimulation in each cycle. When 8 channels were selected, no significant difference was found between the ACE and PACE strategies.

4. DISCUSSION

The results presented suggest that a psychoacoustic model used to select the N bands in "NofM"-type strategies such as ACE can improve speech recognition by cochlear implant subjects in noise. The mean scores for the HSM sentence test were 65% using the psychoacoustic model and 57% for

| D.41 | Number of lists tested | Number of electrodes | Channel stimulation | Noise or quiet |
|------------|------------------------|----------------------|---------------------|----------------|
| Patient 1d | for each condition | selected | rate | conditions |
| P1 | 6 | 4 | 1080 | Noise |
| P2 | 2 | 4 | 1080 | Noise |
| P2 | 2 | 8 | 1080 | Noise |
| Р3 | 4 | 4 | 900 | Noise |
| P4 | 3 | 4 | 500 | Quiet |
| Р5 | 2 | 4 | 900 | Noise |
| Р5 | 2 | 8 | 900 | Noise |
| P6 | 2 | 4 | 900 | Noise |
| P6 | 2 | 8 | 900 | Noise |
| P7 | 2 | 4 | 1200 | Noise |
| P7 | 2 | 8 | 1200 | Noise |
| P8 | 2 | 4 | 900 | Quiet |
| P8 | 2 | 8 | 900 | Quiet |

TABLE 4: Test details for each patient.



FIGURE 12: Score by subject (average and standard deviation). Unless otherwise specified, scores were obtained in noise conditions with a signal-to-noise ratio of 15 dB. The tests were performed using 4 electrodes in each cycle of stimulation.

the ACE strategy when 8 electrodes were stimulated in each cycle of stimulation. The mean score obtained with 4 electrodes stimulated was 67% using the psychoacoustic model and 50% for the ACE strategy. Results were only statistically significant under the 4-channel condition; it is, however, possible that future studies with larger sample sizes may yield significant results for the 8-channel condition as well. Interestingly, performance using PACE was virtually the same regardless of whether 4 or 8 electrodes were used. Therefore, a considerable energy saving could be made using the PACE strategy as it is able to generate the same scores as the ACE strategy while stimulating only half as many electrodes.

Another advantage is that the bands selected using a psychoacoustic model are more widely separated over the frequency domain. It can be speculated that interaction between channels could therefore be reduced. Additionally, the choice of bands is not merely a matter of selecting the largest amplitudes (as with the ACE); this means that smaller electrical currents are required, resulting in power savings.

As can be observed from the results, the difference between the PACE and ACE strategies only exists for the 4channel condition. This may be because selecting fewer channels means that the spectrum of the audio signal is more poorly represented. This places more of a premium on selecting the right signal components. Using a psychoacoustic model to select the bands appears to be a superior approach to just selecting the channels with higher amplitudes—at least, that is, if the number of selected channels is small. As more channels are selected (8-channel condition), the spectrum of the audio signal is better represented and the selection of the most important components becomes less relevant.



FIGURE 13: Score by subject (average and standard deviation). Unless otherwise specified, scores were obtained in noise conditions with a signal-to-noise ratio of 15 dB. The tests were performed with 8 electrodes in each cycle of stimulation.

| TABLE 5: Statistical analysi |
|------------------------------|
|------------------------------|

| Condition 1 | Condition 2 | Test | p-value |
|----------------|----------------|----------|-----------------------------|
| 4-channel ACE | 4-channel PACE | Wilcoxon | Significant $p = 0.017$ |
| 8-channel ACE | 8-channel PACE | Wilcoxon | Not significant $p = 0.138$ |
| 4-channel PACE | 8-channel PACE | Wilcoxon | Not significant $p = 0.786$ |
| 4-channel ACE | 8-channel ACE | Wilcoxon | Significant $p = 0.043$ |

The choice of the parameters that define the spreading function requires more thorough investigation in the future. The spreading function determines how much one channel masks the adjacent frequency bands. As this is not a longterm study and subjects' attention span during speech perception tests is limited, only two different parameter sets are investigated in this paper. The spreading function determined by the first parameter set presented a stronger masking effect than the spreading function determined by the second parameter set. First experiments with cochlear implant subjects revealed that stronger masking effect results in poorer speech perception. One explanation for this might be that important speech cues are being left out by the wider masking curves, which then become inaudible to the subject. Nevertheless, the results obtained thus far are encouraging and indicate the usefulness of a psychoacousticmasking model in the field of cochlear implants. As the optimal parameter set might vary among subjects, further studies are planned to determine the optimal parameter set for the psychoacoustic-masking model. There are also plans to incorporate masking effects whose occurrence may be due

to overlapping of the electrical fields inside the cochlea. The excitation of a subset of neurons that are being stimulated by adjacent electrodes can be determined by measurements using the neural response telemetry (NRT) capabilities of the Nucleus 24 implant [33]. The data derived from such tests can be used to determine the degree of channel interaction [34] and this knowledge could be additionally exploited in a future version of our masking model. There is, however, currently only relatively limited data on electrical masking in cochlear implant subjects, and this influenced the authors' decision to initially concentrate on a psychoacoustic-masking model for which fundamental knowledge was already available.

It should be reiterated that our research ACE strategy and the new PACE strategy used for the tests do not make use of a pre-emphasis filter. The ACE and PACE strategies process signals fed directly from a computer hard disk, so that the analogue front end of the speech processor containing both pre-emphasis and AGC functionality is bypassed. The high-frequency gain usually leads to the ACE strategy selecting higher-frequency bands than when a pre-emphasis filter is absent, and high-frequency components are important for speech understanding. The PACE strategy may already account partially for the lack of pre-emphasis by introducing the absolute threshold in quiet function where the higher-frequency parts of a white-noise signal are more above threshold than the low-frequency parts. For this reason the effect of the pre-emphasis may work differently for the PACE strategy than for the ACE strategy.

Another important aspect is the complexity of the new PACE strategy. As presented in Section 3, this strategy uses the same block structure as the ACE strategy but incorporates a psychoacoustic model to select the bands. This allowed the major blocks of the ACE strategy to be adopted for the PACE strategy. Our implementation of PACE on a personal computer was not specifically optimized in terms of computational efficiency. However, it is worth mentioning that the PACE strategy has been already implemented in a commercial speech processor for chronic investigations not posing any challenge in terms of computational demands.

The selection of the appropriate signal components is obviously of great importance. The introduction of simple "NofM" approaches in the 1990s already represented a significant improvement over conventional CIS-like strategies by stimulating fewer electrodes per frame but increasing the channel rate in each channel [6, 7, 35]. However, the stimulation rate may not be the only factor contributing to better hearing with "NofM"-type strategies, as researchers have also observed that these strategies have advantages over CIS-like speech coding using comparable stimulation rates [6, 7, 8]. The close relationship between "NofM"-type strategies and psychoacoustic masking has been already been mentioned in [35].

Advances in the field of speech coding mean that understanding in quiet is no longer a major problem for most recipients, although hearing in noisy conditions is still severely limited [36, 37]. Nevertheless, technical progress in this field has in the recent past led to remarkable performance enhancements in device users. Moreover, intelligent new speech coding strategies such as transient emphasis spectral maxima (TESM), which emphasize certain cues in the audio signal, have demonstrated improvement in terms of speech perception [38]. However, the electrode-nerve interface that is intended to substitute for the hair cells inside the cochlea is clearly not remotely as sophisticated as a fully functional cochlea. With today's systems we are attempting to mimic thousands of nerve fibres using crude electrode arrays that contain 8 to 22 electrode contacts at most. Bearing these limitations in mind, it becomes apparent that the way in which these few electrodes are selected and stimulated plays a key role in helping cochlear implant subjects understand speech in difficult hearing situations.

5. CONCLUSIONS

The results of the PACE strategy, as described above, suggest that psychoacoustic masking is also applicable to cochlear implant recipients. The idea behind the PACE strategy was to present to users of such devices only those signal components that are most clearly perceived by normal-hearing people. In so doing, the limited resolution of the cochlear implant and the electrode-nerve interface can be used more effectively. Results obtained with device users showed significant improvement in speech perception when 4 electrodes were selected using the PACE strategy. No significant improvement was found when 8 electrodes were selected.

One important final comment: it can be expected that the adoption of a psychoacoustic model in speech processors for chronic use may result in even higher scores using the new PACE strategy. The implementation of a psychoacoustic model increases the complexity of simpler "NofM" approaches. However, its implementation is clearly viable in commercial speech processor for cochlear implants. We are currently setting up a long-term study on the PACE strategy which will be conducted using a commercially available speech processor, thus utilizing the usual analogue front end (with AGC and pre-emphasis filter) and giving users takehome experience.

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In the case of vision, the stimuli go through a complex analysis chain along the so-called "visual pathway," starting with the encoding by the photoreceptors in the retina (low-level processing) and ending with cognitive mechanisms (high-level processes) that depend on the task being performed.

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Call for Papers

The main focus of this special issue is on the application of digital signal processing techniques for music information retrieval (MIR). MIR is an emerging and exciting area of research that seeks to solve a wide variety of problems dealing with preserving, analyzing, indexing, searching, and accessing large collections of digitized music. There are also strong interests in this field of research from music libraries and the recording industry as they move towards digital music distribution. The demands from the general public for easy access to these music libraries challenge researchers to create tools and algorithms that are robust, small, and fast.

Music is represented in either encoded audio waveforms (CD audio, MP3, etc.) or symbolic forms (musical score, MIDI, etc.). Audio representations, in particular, require robust signal processing techniques for many applications of MIR since meaningful descriptions need to be extracted from audio signals in which sounds from multiple instruments and vocals are often mixed together. Researchers in MIR are therefore developing a wide range of new methods based on statistical pattern recognition, classification, and machine learning techniques such as the Hidden Markov Model (HMM), maximum likelihood estimation, and Bayes estimation as well as digital signal processing techniques such as Fourier and Wavelet transforms, adaptive filtering, and source-filter models. New music interface and query systems leveraging such methods are also important for end users to benefit from MIR research.

Although research contributions on MIR have been published at various conferences in 1990s, the members of the MIR research community meet annually at the International Conference on Music Information Retrieval (ISMIR) since 2000.

Topics of interest include (but are not limited to):

- Automatic summarization (succinct representation of music)
- Automatic transcription (audio to symbolic format conversion)
- Music annotation (semantic analysis)
- Music fingerprinting (unique identification of music)
- Music interface
- Music similarity metrics (comparison)

- Music understanding
- Musical feature extraction
- Musical styles and genres
- Optical music score recognition (image to symbolic format conversion)
- Performer/artist identification
- Query systems
- Timbre/instrument recognition

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Special Issue on Visual Sensor Networks

Call for Papers

Research into the design, development, and deployment of networked sensing devices for high-level inference and surveillance of the physical environment has grown tremendously in the last few years.

This trend has been motivated, in part, by recent technological advances in electronics, communication networking, and signal processing.

Sensor networks are commonly comprised of lightweight distributed sensor nodes such as low-cost video cameras. There is inherent redundancy in the number of nodes deployed and corresponding networking topology. Operation of the network requires autonomous peer-based collaboration amongst the nodes and intermediate data-centric processing amongst local sensors. The intermediate processing known as in-network processing is application-specific. Often, the sensors are untethered so that they must communicate wirelessly and be battery-powered. Initial focus was placed on the design of sensor networks in which scalar phenomena such as temperature, pressure, or humidity were measured.

It is envisioned that much societal use of sensor networks will also be based on employing content-rich vision-based sensors. The volume of data collected as well as the sophistication of the necessary in-network stream content processing provide a diverse set of challenges in comparison with generic scalar sensor network research.

Applications that will be facilitated through the development of visual sensor networking technology include automatic tracking, monitoring and signaling of intruders within a physical area, assisted living for the elderly or physically disabled, environmental monitoring, and command and control of unmanned vehicles.

Many current video-based surveillance systems have centralized architectures that collect all visual data at a central location for storage or real-time interpretation by a human operator. The use of distributed processing for automated event detection would significantly alleviate mundane or time-critical activities performed by human operators, and provide better network scalability. Thus, it is expected that video surveillance solutions of the future will successfully utilize visual sensor networking technologies. Given that the field of visual sensor networking is still in its infancy, it is critical that researchers from the diverse disciplines including signal processing, communications, and electronics address the many challenges of this emerging field. This special issue aims to bring together a diverse set of research results that are essential for the development of robust and practical visual sensor networks.

Topics of interest include (but are not limited to):

- Sensor network architectures for high-bandwidth vision applications
- Communication networking protocols specific to visual sensor networks
- Scalability, reliability, and modeling issues of visual sensor networks
- Distributed computer vision and aggregation algorithms for low-power surveillance applications
- Fusion of information from visual and other modalities of sensors
- Storage and retrieval of sensor information
- Security issues for visual sensor networks
- Visual sensor network testbed research
- Novel applications of visual sensor networks
- Design of visual sensors

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Special Issue on Multirate Systems and Applications

Call for Papers

Filter banks for the application of subband coding of speech were introduced in the 1970s. Since then, filter banks and multirate systems have been studied extensively. There has been great success in applying multirate systems to many applications. The most notable of these applications include subband coding for audio, image, and video, signal analysis and representation using wavelets, subband denoising, and so forth. Different applications also call for different filter bank designs and the topic of designing one-dimensional and multidimentional filter banks for specific applications has been of great interest.

Recently there has been growing interest in applying multirate theories to the area of communication systems such as, transmultiplexers, filter bank transceivers, blind deconvolution, and precoded systems. There are strikingly many dualities and similarities between multirate systems and multicarrier communication systems. Many problems in multicarrier transmission can be solved by extending results from multirate systems and filter banks. This exciting research area is one that is of increasing importance.

The aim of this special issue is to bring forward recent developments on filter banks and the ever-expanding area of applications of multirate systems.

Topics of interest include (but are not limited to):

- Multirate signal processing for communications
- Filter bank transceivers
- One-dimensional and multidimensional filter bank designs for specific applications
- Denoising
- Adaptive filtering
- Subband coding
- Audio, image, and video compression
- Signal analysis and representation
- Feature extraction and classification
- Other applications

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Special Issue on

Multisensor Processing for Signal Extraction and Applications

Call for Papers

Source signal extraction from heterogeneous measurements has a wide range of applications in many scientific and technological fields, for example, telecommunications, speech and acoustic signal processing, and biomedical pattern analysis. Multiple signal reception through multisensor systems has become an effective means for signal extraction due to its superior performance over the monosensor mode. Despite the rapid progress made in multisensor-based techniques in the past few decades, they continue to evolve as key technologies in modern wireless communications and biomedical signal processing. This has led to an increased focus by the signal processing community on the advanced multisensorbased techniques which can offer robust high-quality signal extraction under realistic assumptions and with minimal computational complexity. However, many challenging tasks remain unresolved and merit further rigorous studies. Major efforts in developing advanced multisensor-based techniques may include high-quality signal extraction, realistic theoretical modeling of real-world problems, algorithm complexity reduction, and efficient real-time implementation.

The purpose of this special issue aims to present state-ofthe-art multisensor signal extraction techniques and applications. Contributions in theoretical study, performance analysis, complexity reduction, computational advances, and realworld applications are strongly encouraged.

Topics of interest include (but are not limited to):

- Multiantenna processing for radio signal extraction
- Multimicrophone speech recognition and enhancement
- Multisensor radar, sonar, navigation, and biomedical signal processing
- Blind techniques for multisensor signal extraction
- Computational advances in multisensor processing

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Special Issue on

Search and Retrieval of 3D Content and Associated Knowledge Extraction and Propagation

Call for Papers

With the general availability of 3D digitizers, scanners, and the technology innovation in 3D graphics and computational equipment, large collections of 3D graphical models can be readily built up for different applications (e.g., in CAD/CAM, games design, computer animations, manufacturing and molecular biology). For such large databases, the method whereby 3D models are sought merits careful consideration. The simple and efficient query-by-content approach has, up to now, been almost universally adopted in the literature. Any such method, however, must first deal with the proper positioning of the 3D models. The two prevalent-in-the-literature methods for the solution to this problem seek either

- Pose Normalization: Models are first placed into a canonical coordinate frame (normalizing for translation, scaling, and rotation). Then, the best measure of similarity is found by comparing the extracted feature vectors, or
- Descriptor Invariance: Models are described in a transformation invariant manner, so that any transformation of a model will be described in the same way, and the best measure of similarity is obtained at any transformation.

The existing 3D retrieval systems allow the user to perform queries by example. The queried 3D model is then processed, low-level geometrical features are extracted, and similar objects are retrieved from a local database. A shortcoming of the methods that have been proposed so far regarding the 3D object retrieval, is that neither is the semantic information (high-level features) attached to the (low-level) geometric features of the 3D content, nor are the personalization options taken into account, which would significantly improve the retrieved results. Moreover, few systems exist so far to take into account *annotation* and *relevance feedback* techniques, which are very popular among the corresponding content-based image retrieval systems (CBIR).

Most existing CBIR systems using knowledge either annotate all the objects in the database (full annotation) or annotate a subset of the database manually selected (partial annotation). As the database becomes larger, full annotation is increasingly difficult because of the manual effort needed. Partial annotation is relatively affordable and trims down the heavy manual labor. Once the database is partially annotated, traditional image analysis methods are used to derive semantics of the objects not yet annotated. However, it is not clear "how much" annotation is sufficient for a specific database and what the best subset of objects to annotate is. In other words how the knowledge *will be propagated*. Such techniques have not been presented so far regarding the 3D case.

Relevance feedback was first proposed as an interactive tool in text-based retrieval. Since then it has been proven to be a powerful tool and has become a major focus of research in the area of content-based search and retrieval. In the traditional computer centric approaches, which have been proposed so far, the "best" representations and weights are fixed and they cannot effectively model high-level concepts and user's perception subjectivity. In order to overcome these limitations of the computer centric approach, techniques based on *relevant feedback*, in which the human and computer interact to refine high-level queries to representations based on low-level features, should be developed.

The aim of this special issue is to focus on recent developments in this expanding research area. The special issue will focus on novel approaches in 3D object retrieval, transforms and methods for efficient geometric feature extraction, annotation and relevance feedback techniques, knowledge propagation (e.g., using Bayesian networks), and their combinations so as to produce a single, powerful, and dominant solution.

Topics of interest include (but are not limited to):

- 3D content-based search and retrieval methods (volume/surface-based)
- Partial matching of 3D objects
- Rotation invariant feature extraction methods for 3D objects

- Graph-based and topology-based methods
- 3D data and knowledge representation
- Semantic and knowledge propagation over heterogeneous metadata types
- Annotation and relevance feedback techniques for 3D objects

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Special Issue on Robust Speech Recognition

Call for Papers

Robustness can be defined as the ability of a system to maintain performance or degrade gracefully when exposed to conditions not well represented in the data used to develop the system. In automatic speech recognition (ASR), systems must be robust to many forms of signal degradation, including speaker characteristics (e.g., dialect and accent), ambient environment (e.g., cellular telephony), transmission channel (e.g., voice over IP), and language (e.g., new words, dialect switching). Robust ASR systems, which have been under development for the past 35 years, have made great progress over the years closing the gap between performance on pristine research tasks and noisy operational data.

However, in recent years, demand is emerging for a new class of systems that tolerate extreme and unpredictable variations in operating conditions. For example, in a cellular telephony environment, there are many nonstationary forms of noise (e.g., multiple speakers) and significant variations in microphone type, position, and placement. Harsh ambient conditions typical in automotive and mobile applications pose similar challenges. Development of systems in a language or dialect for which there is limited or no training data in a target language has become a critical issue for a new generation of voice mining applications. The existence of multiple conditions in a single stream, a situation common to broadcast news applications, and that often involves unpredictable changes in speaker, topic, dialect, or language, is another form of robustness that has gained attention in recent years.

Statistical methods have dominated the field since the early 1980s. Such systems tend to excel at learning the characteristics of large databases that represent good models of the operational conditions and do not generalize well to new environments.

This special issue will focus on recent developments in this key research area. Topics of interest include (but are not limited to):

- Channel and microphone normalization
- Stationary and nonstationary noise modeling, compensation, and/or rejection
- Localization and separation of sound sources (including speaker segregation)

- Signal processing and feature extraction for applications involving hands-free microphones
- Noise robust speech modeling
- Adaptive training techniques
- Rapid adaptation and learning
- Integration of confidence scoring, metadata, and other alternative information sources
- Audio-visual fusion
- Assessment relative to human performance
- Machine learning algorithms for robustness
- Transmission robustness
- Pronunciation modeling

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Special Issue on

Signal Processing Technologies for Ambient Intelligence in Home-Care Applications

Call for Papers

The possibility of allowing elderly people with different kinds of disabilities to conduct a normal life at home and achieve a more effective inclusion in the society is attracting more and more interest from both industrial and governmental bodies (hospitals, healthcare institutions, and social institutions). Ambient intelligence technologies, supported by adequate networks of sensors and actuators, as well as by suitable processing and communication technologies, could enable such an ambitious objective.

Recent researches demonstrated the possibility of providing constant monitoring of environmental and biomedical parameters, and the possibility to autonomously originate alarms, provide primary healthcare services, activate emergency calls, and rescue operations through distributed assistance infrastructures. Nevertheless, several technological challenges are still connected with these applications, ranging from the development of enabling technologies (hardware and software), to the standardization of interfaces, the development of intuitive and ergonomic human-machine interfaces, and the integration of complex systems in a highly multidisciplinary environment.

The objective of this special issue is to collect the most significant contributions and visions coming from both academic and applied research bodies working in this stimulating research field. This is a highly interdisciplinary field comprising many areas, such as signal processing, image processing, computer vision, sensor fusion, machine learning, pattern recognition, biomedical signal processing, multimedia, human-computer interfaces, and networking.

The focus will be primarily on the presentation of original and unpublished works dealing with ambient intelligence and domotic technologies that can enable the provision of advanced homecare services.

This special issue will focus on recent developments in this key research area. Topics of interest include (but are not limited to):

- Video-based monitoring of domestic environments and users
- Continuous versus event-driven monitoring
- Distributed information processing

- Data fusion techniques for event association and automatic alarm generation
- Modeling, detection, and learning of user habits for automatic detection of anomalous behaviors
- Integration of biomedical and behavioral data
- Posture and gait recognition and classification
- Interactive multimedia communications for remote assistance
- Content-based encoding of medical and behavioral data
- Networking support for remote healthcare
- Intelligent/natural man-machine interaction, personalization, and user acceptance

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Special Issue on Spatial Sound and Virtual Acoustics

Call for Papers

Spatial sound reproduction has become widespread in the form of multichannel audio, particularly through home theater systems. Reproduction systems from binaural (by headphones) to hundreds of loudspeaker channels (such as wave field synthesis) are entering practical use. The application potential of spatial sound is much wider than multichannel sound, however, and research in the field is active. Spatial sound covers for example the capturing, analysis, coding, synthesis, reproduction, and perception of spatial aspects in audio and acoustics.

In addition to the topics mentioned above, research in virtual acoustics broadens the field. Virtual acoustics includes techniques and methods to create realistic percepts of sound sources and acoustic environments that do not exist naturally but are rendered by advanced reproduction systems using loudspeakers or headphones. Augmented acoustic and audio environments contain both real and virtual acoustic components.

Spatial sound and virtual acoustics are among the major research and application areas in audio signal processing. Topics of active study range from new basic research ideas to improvement of existing applications. Understanding of spatial sound perception by humans is also an important area, in fact a prerequisite to advanced forms of spatial sound and virtual acoustics technology.

This special issue will focus on recent developments in this key research area. Topics of interest include (but are not limited to):

- Multichannel reproduction
- Wave field synthesis
- Binaural reproduction
- Format conversion and enhancement of spatial sound
- Spatial sound recording
- Analysis, synthesis, and coding of spatial sound
- Spatial sound perception and auditory modeling
- Simulation and modeling of room acoustics
- Auralization techniques
- Beamforming and sound source localization
- Acoustic and auditory scene analysis
- Augmented reality audio

- Virtual acoustics (sound environments and sources)
- Intelligent audio environments
- Loudspeaker-room interaction and equalization
- Applications

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Special Issue on

Advances in Electrocardiogram Signal Processing and Analysis

Call for Papers

Since its invention in the 19th century when it was little more than a scientific curiosity, the electrocardiogram (ECG) has developed into one of the most important and widely used quantitative diagnostic tools in medicine. It is essential for the identification of disorders of the cardiac rhythm, extremely useful for the diagnosis and management of heart abnormalities such as myocardial infarction (heart attack), and offers helpful clues to the presence of generalised disorders that affect the rest of the body, such as electrolyte disturbances and drug intoxication.

Recording and analysis of the ECG now involves a considerable amount of signal processing for S/N enhancement, beat detection, automated classification, and compression. These involve a whole variety of innovative signal processing methods, including adaptive techniques, time-frequency and time-scale procedures, artificial neural networks and fuzzy logic, higher-order statistics and nonlinear schemes, fractals, hierarchical trees, Bayesian approaches, and parametric models, amongst others.

This special issue will review the current status of ECG signal processing and analysis, with particular regard to recent innovations. It will report major achievements of academic and commercial research institutions and individuals, and provide an insight into future developments within this exciting and challenging area.

This special issue will focus on recent developments in this key research area. Topics of interest include (but are not limited to):

- Beat (QRS complex) detection
- ECG compression
- Denoising of ECG signals
- Morphological studies and classification
- ECG modeling techniques
- Expert systems and automated diagnosis
- QT interval measurement and heart-rate variability
- Arrhythmia and ischemia detection and analysis
- Interaction between cardiovascular signals (ECG, blood pressure, respiration, etc.)

- Intracardiac ECG analysis (implantable cardiovascular devices, and pacemakers)
- ECGs and sleep apnoea
- Real-time processing and instrumentation
- ECG telemedicine and e-medicine
- Fetal ECG detection and analysis
- Computational tools and databases for ECG education and research

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Special Issue on

Emerging Signal Processing Techniques for Power Quality Applications

Call for Papers

Recently, end users and utility companies are increasingly concerned with perturbations originated from electrical power quality variations. Investigations are being carried out to completely characterize not only the old traditional type of problems, but also new ones that have arisen as a result of massive use of nonlinear loads and electronics-based equipment in residences, commercial centers, and industrial plants. These nonlinear load effects are aggravated by massive power system interconnections, increasing number of different power sources, and climatic changes.

In order to improve the capability of equipments applied to monitoring the power quality of transmission and distribution power lines, power systems have been facing new analysis and synthesis paradigms, mostly supported by signal processing techniques. The analysis and synthesis of emerging power quality and power system problems led to new research frontiers for the signal processing community, focused on the development and combination of computational intelligence, source coding, pattern recognition, multirate systems, statistical estimation, adaptive signal processing, and other digital processing techniques, implemented in either DSP-based, PC-based, or FPGA-based solutions.

The goal of this proposal is to introduce powerful and efficient real-time or almost-real-time signal processing tools for dealing with the emerging power quality problems. These techniques take into account power-line signals and complementary information, such as climatic changes.

This special issue will focus on recent developments in this key research area. Topics of interest include (but are not limited to):

- Detection of transients
- Classification of multiple events
- Identification of isolated and multiple disturbance sources
- Compression of voltage and current data signals
- Location of disturbance sources
- Prediction of transmission and distribution systems failures
- Demand forecasting

• Parameters estimation for fundamental, harmonics, and interharmonics

Digital signal processing techniques applied to power quality applications are a very attractive and stimulating area of research. Its results will provide, in the near future, new standards for the decentralized and real-time monitoring of transmission and distribution systems, allowing to closely follow and predict power system performance. As a result, the power systems will be more easily planned, expanded, controlled, managed, and supervised.

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NEWS RELEASE Nominations Invited for the Institute of Acoustics 2006 A B Wood Medal

The Institute of Acoustics, the UK's leading professional body for those working in acoustics, noise and vibration, is inviting nominations for its prestigious A B Wood Medal for the year 2006.

The A B Wood Medal and prize is presented to an individual, usually under the age of 35, for distinguished contributions to the application of underwater acoustics. The award is made annually, in even numbered years to a person from Europe and in odd numbered years to someone from the USA/Canada. The 2005 Medal was awarded to Dr A Thode from the USA for his innovative, interdisciplinary research in ocean and marine mammal acoustics.

Nominations should consist of the candidate's CV, clearly identifying peer reviewed publications, and a letter of endorsement from the nominator identifying the contribution the candidate has made to underwater acoustics. In addition, there should be a further reference from a person involved in underwater acoustics and not closely associated with the candidate. Nominees should be citizens of a European Union country for the 2006 Medal. Nominations should be marked confidential and addressed to the President of the Institute of Acoustics at 77A St Peter's Street, St. Albans, Herts, AL1 3BN. The deadline for receipt of nominations is **15 October 2005**.

Dr Tony Jones, President of the Institute of Acoustics, comments, "A B Wood was a modest man who took delight in helping his younger colleagues. It is therefore appropriate that this prestigious award should be designed to recognise the contributions of young acousticians."

> Further information and an nomination form can be found on the Institute's website at www.ioa.org.uk.

A B Wood

Albert Beaumont Wood was born in Yorkshire in 1890 and graduated from Manchester University in 1912. He became one of the first two research scientists at the Admiralty to work on antisubmarine defence. He designed the first directional hydrophone and was well known for the many contributions he made to the science of underwater acoustics and for the help he gave to younger colleagues. The medal was instituted after his death by his many friends on both sides of the Atlantic and was administered by the Institute of Physics until the formation of the Institute of Acoustics in 1974.

PRESS CONTACT

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EDITORS NOTES

The Institute of Acoustics is the UK's professional body for those working in acoustics, noise and vibration. It was formed in 1974 from the amalgamation of the Acoustics Group of the Institute of Physics and the British Acoustical Society (a daughter society of the Institution of Mechanical Engineers). The Institute of Acoustics is a nominated body of the Engineering Council, offering registration at Chartered and Incorporated Engineer levels.

The Institute has some 2500 members from a rich diversity of backgrounds, with engineers, scientists, educators, lawyers, occupational hygienists, architects and environmental health officers among their number. This multidisciplinary culture provides a productive environment for cross-fertilisation of ideas and initiatives. The range of interests of members within the world of acoustics is equally wide, embracing such aspects as aerodynamics, architectural acoustics, building acoustics, electroacoustics, engineering dynamics, noise and vibration, hearing, speech, underwater acoustics, together with a variety of environmental aspects. The lively nature of the Institute is demonstrated by the breadth of its learned society programmes.

For more information please visit our site at www.ioa.org.uk.

EURASIP Book Series on SP&C, Volume 1, ISBN 977-5945-08-9

HIGH-FIDELITY MULTICHANNEL AUDIO CODING

Dai Tracy Yang, Chris Kyriakakis, and C.-C. Jay Kuo



This invaluable monograph addresses the specific needs of audio-engineering students and researchers who are either learning about the topic or using it as a reference book on multichannel audio compression. This book covers a wide range of knowledge on perceptual audio coding, from basic digital signal processing and data compression techniques to advanced audio coding standards and innovative coding tools. It is the only book available on the market that solely focuses on the principles of highquality audio codec design for multichannel sound sources.

This book includes three parts. The first part covers the basic topics on audio compression, such as quantization, entropy coding, psychoacoustic model, and sound quality assessment. The second part of the book highlights the current most prevalent low-bit-rate high-performance audio coding standards—MPEG-4 audio. More space is given to the audio standards that are capable of supporting multichannel signals, that is, MPEG advanced audio coding (AAC), including the original MPEG-2 AAC technology, additional MPEG-4 toolsets, and the most recent aacPlus standard. The third part of this book introduces several innovative multichannel audio coding

tools, which have been demonstrated to further improve the coding performance and expand the available functionalities of MPEG AAC, and is more suitable for graduate students and researchers in the advanced level.

Dai Tracy Yang is currently Postdoctoral Research Fellow, Chris Kyriakakis is Associated Professor, and C.-C. Jay Kuo is Professor, all affiliated with the Integrated Media Systems Center (IMSC) at the University of Southern California.

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EURASIP Book Series on SP&C, Volume 2, ISBN 977-5945-07-0

GENOMIC SIGNAL PROCESSING AND STATISTICS

Edited by: Edward R. Dougherty, Ilya Shmulevich, Jie Chen, and Z. Jane Wang



Recent advances in genomic studies have stimulated synergetic research and development in many cross-disciplinary areas. Genomic data, especially the recent large-scale microarray gene expression data, represents enormous challenges for signal processing and statistics in processing these vast data to reveal the complex biological functionality. This perspective naturally leads to a new field, genomic signal processing (GSP), which studies the processing of genomic signals by integrating the theory of signal processing and statistics. Written by an international, interdisciplinary team of authors, this invaluable edited volume is accessible to students just entering this emergent field, and to researchers, both in academia and industry, in the fields of molecular biology, engineering, statistics, and signal processing. The book provides tutorial-level overviews and addresses the specific needs of genomic signal processing students and researchers as a reference book.

The book aims to address current genomic challenges by exploiting potential synergies between genomics, signal processing, and statistics, with special emphasis on signal processing and statistical tools for

structural and functional understanding of genomic data. The book is partitioned into three parts. In part I, a brief history of genomic research and a background introduction from both biological and signal-processing/ statistical perspectives are provided so that readers can easily follow the material presented in the rest of the book. In part II, overviews of state-of-the-art techniques are provided. We start with a chapter on sequence analysis, and follow with chapters on feature selection, clustering, and classification of microarray data. The next three chapters discuss the modeling, analysis, and simulation of biological regulatory networks, especially gene regulatory networks based on Boolean and Bayesian approaches. The next two chapters treat visualization and compression of gene data, and supercomputer implementation of genomic signal processing systems. Part II concludes with two chapters on systems biology and medical implications of genomic research. Finally, part III discusses the future trends in genomic signal processing and statistics research.

For more information and online orders please visit: http://www.hindawi.com/books/spc/volume-2/ For any inquiries on how to order this title please contact books.orders@hindawi.com

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